

Chapter Three

Equivalent Circuit of IM

Losses in I.M are three types ;

i-Iron losses (magnetic losses)

ii-Copper losses (Electrical losses)

iii-Mechanical losses ($f + w$ (friction + windage))losses, or rotational losses.

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Iron losses:

The presence of rotating magnetic field in the stator core account for major portion of the magnetic losses. As the stator core is made of silicon content steel-iron stampings ,also which is called core losses, iron losses consist of

1-Hystereas losses P_h

2-Eddy current losses P_e

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Hysteresis losses:

Every part of stator core is subjected to alternate changes in the magnetic field due to its rotating nature. The frequency of flux reversal (f) is similar to the supply frequency.

The flux reversals take place through the B-H loop or the hysteresis loop causing hysteresis which is given by,

$$P_h = K_h \times B_m^x \times f \quad \text{in watt}$$

Where B_m = Maximum flux density in Tesla in stator core,

f = stator supply frequency,

x = Variable depends on the quality of the material which is stator core stamping are made of.

K_h = Hysteresis constant also depends of materials of stator core.

If the silicon is added to stator core in the manufacture of core the value of K_h is decrease, and then the hysteresis losses is decreases.

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Eddy Current Losses:

The stator core steel may be account as a good medium for magnetic field. Therefore, according to Farady's law the flux reversal in the stator core causes an emf which is called eddy emf induced in stator core. The value of this emf according to this law proportional to flux density and the frequency of flux reversal f . This emf small in value, drives a current called eddy current loss.

To decrease the eddy current losses, the core resistance must be as large as possible. This is achieved by building the stator core with thin laminations, stacked and riveted at right angles to the path of eddy current. The laminations are insulated from each other by a thin coat of varnish.

$$P_e = K_e B_m^2 f^2 \quad \text{in watt}$$

Where K_e = eddy current constant, proportional to the squire of thickness of core lamination and is inversely proportional to the resistively of the core material.

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Iron Losses

$$*P_{iron} = P_h + P_e*$$

The frequency of supply f and flux density B_m is constant, hence P_i may be considered as constant under all conditions. Since the frequency of rotor is very small, the iron losses rotor can be neglected at lower slip (high speed).

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Copper Losses

It is occur due to ohmic resistance of stator and rotor resistance of three phase windings.

$$\begin{aligned} \text{Copper loss} &= \text{stator cu loss} + \text{rotor cu loss} \\ &= 3I_1^2 R_1 + 3 I_2^2 R_2 \end{aligned}$$

Since the I_1 & I_2 vary with load, therefore the cu loss vary with square proportion of the current. Hence cu losses or electrical loss can otherwise be called as variable losses.

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Mechanical Losses (P_m):

Friction Loss (P_f):

Which is occur between solid parts of the machines (friction of bearing). In WRIM the slip ring friction loss added to bearing loss, therefore the WRIM has more friction loss than squirrel cage type.

$$P_f = K_h n \quad (\text{Watt})$$

Where n = rotor speed and K_h is friction loss constant.

Windage Loss(P_w):

This loss occur between a rotating solid part of the machine and the air around the surface of the rotor. Addition windage losses may be added if fan blades are fitted on one side of the rotor for ventilation purposes.

$$P_w = K_w n^2 \quad (\text{Watt})$$

Where K_w = windage loss constant.

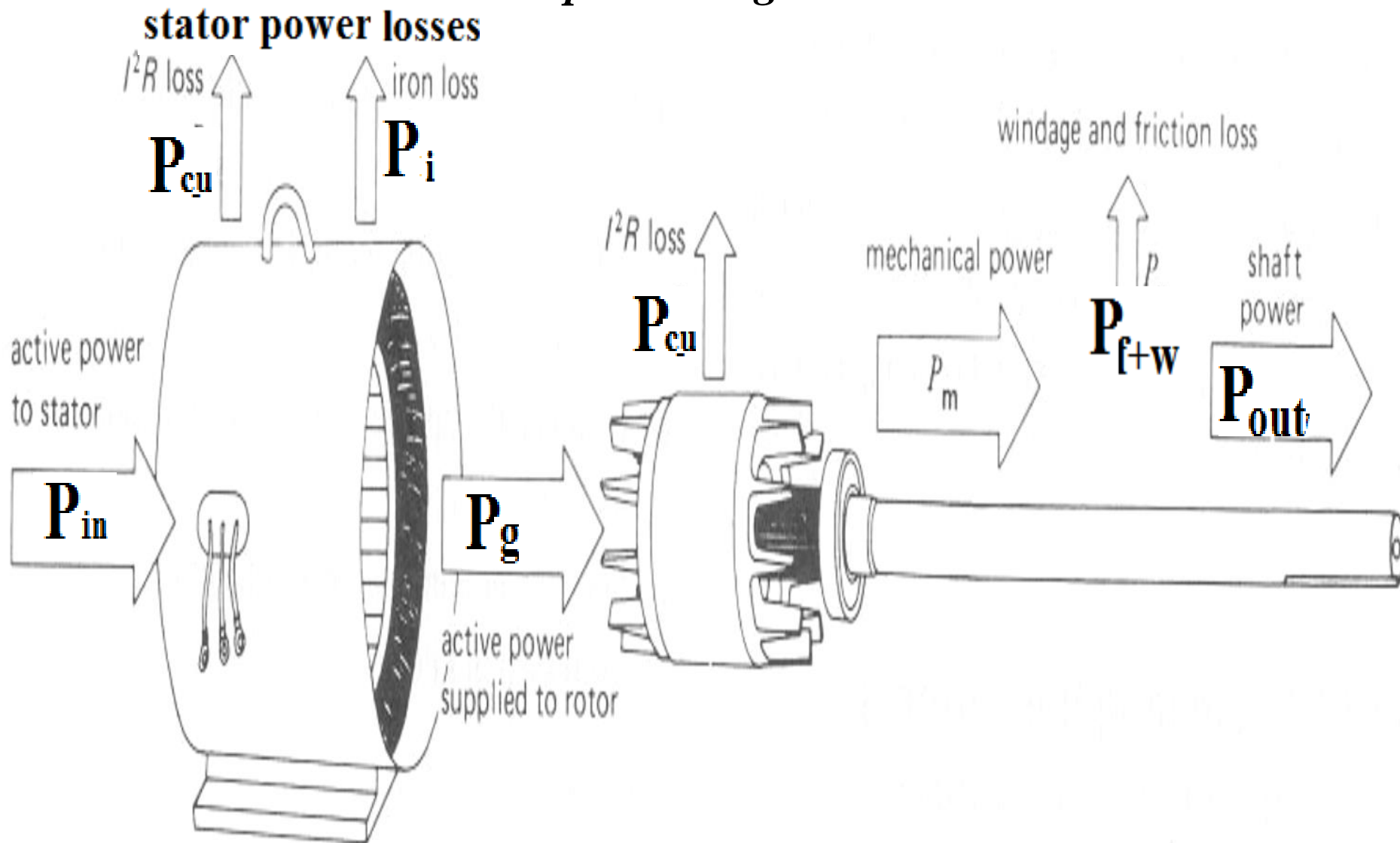
The total mechanical losses $P_m = P_f + P_w = P_{f+w}$

The change in speed from no-load to full load speed is very small, therefore , the mechanical losses are considered to be constant losses as similar to iron losses.

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The various power stages shown below.



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$$P_{\text{inp}} = \sqrt{3} V_L I_L \cos \phi$$

= rotor input + stator cu losses

$$P_g = \text{rotor input} = \text{air gap power}$$
$$= E_1 I_2' \cos \phi_2$$

The air gap power P_g has two parts:

1- $(I_2')^2 R_2' \Rightarrow$ It is converted to electrical power and its dissipated as a heat losses.

2- $(1-S) (I_2')^2 R_2' / S \Rightarrow$ It is converted to mechanical energy

$$P_g / \text{phase} = (I_2')^2 R_2' / S$$
$$= (I_2')^2 R_2' + (1-S) (I_2')^2 R_2' / S$$

= Rotor cu loss (P_{cu}) + mechanical loss (p_{mec})

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$$P_g \text{ /phase} = S P_g + P_g (1-S)$$

The rotor output is converted into mechanical energy. some is lost due to windage and friction losses in the rotor and the rest appears as the useful or shaft torque T_{sh} or T_{out} .

$$P_g = T_{airgap} \cdot \omega_s$$
$$= 2\pi n_s \cdot T_e$$

Te it is called air gap torque or gross torque.

$$T_e = \frac{P_g}{\omega_s}$$

$$\frac{\text{Rotorcu loss}}{\text{Rotorinput}} = \frac{SP_g}{P_g} = S = \frac{n_s - n}{n_s}$$

$$\text{Rotorcu loss} = Sp_g$$

$$\text{Rotor grossoutput} = P_g (1 - S)$$

$$\frac{\text{Rotoroutput}(P_{mech})}{\text{Rotor Input}(P_g)} = \frac{P_g (1 - S)}{P_g} = 1 - S = 1 - \frac{n_s - n}{n_s}$$

$$\frac{P_{mech}}{P_g} = \frac{n}{n_s}$$

$$\text{Rotor efficiency} = \frac{P_{mech}}{P_g} = \frac{n}{n_s}$$

$$T_{out} = \frac{P_{mech}}{\omega_m} = \frac{P_{mech}}{2\pi n_r}$$

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Example:

The power input to the rotor of 415V, 50Hz, 4-pole, three phase IM is 55kW, the frequency of rotor is 2Hz, find, the slip, the rotor speed (n), the rotor cu losses and mechanical power developed.

Solution:

$$n_s = 120f/p = 1500 \text{rpm}$$

$$f_r = Sf \quad \Rightarrow S = f_r/f = 2/50 = 0.04$$

$$0.04 = \frac{1500 - n}{1500}, \quad \therefore n = 1440 \text{rpm}$$

$$\text{Rotor cu loss } P_{cu} = SP_g = 0.04 * 55 \text{kW} = 2.2 \text{kW}$$

$$P_{mech} = P_g(1-S) = 55 * 10^3 (1-0.04) = 52.8 \text{kW}$$

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Example: A 415V, 4-pole, three phase IM develops 15.25kW with mechanical losses when running at 1426rpm. The supply frequency 50.5Hz with power factor of 0.89. find; the speed, rotor copper losses, input power to stator if stator losses are 1.4kW, input line current and rotor frequency .

Solution:

$$n_s = \frac{120 \times 50.5}{4} = 1515 \text{rpm}, \quad S = \frac{1515 - 1462}{1515} = 0.035$$

$$P_{mech} = P_s (1 - S) \Rightarrow P_s = \frac{P_{mech}}{1 - S} = \frac{15.25 \times 10^3}{1 - 0.035} = 15803 \text{Watt}$$

$$CU_r = SP_s = 0.035 \times 15803 = 553 \text{Watt}$$

$$P_{in} = P_s + \text{stator loss} = 15803 + 1400 = 17.203 \text{kWatt}$$

$$P_{in} = \sqrt{3} V_L I_L \cos \phi \Rightarrow 17203.1 = \sqrt{3} \times 415 \times I_L \times 0.89 \Rightarrow$$

$$I_L = 26.891 \text{A}, \quad \& f_r = 0.035 \times 50.4 = 1.76 \text{Hz}$$

Example:

A 4-pole, 415V, 50Hz, three phase IM deliver a shaft useful torque of 25.04N.m with a speed of 1411rpm when the supply frequency is 49Hz. The stator losses equal to 320Watt and input power factor is 0.86. If the mechanical torque lost in (f+w) is 0.5N.m. find; the input current and efficiency.

Solution:

$$n_s = \frac{120 \times 49}{4} = 1470 \text{rpm}, \quad S = \frac{1470 - 1411}{1470} = 0.04$$

$$T_{\text{mech}} = 25.04 \text{N.m} \Rightarrow P_{\text{mech}} = \omega_{\text{mech}} T_{\text{mech}} = 147.76 \times 25.04 = 3700 \text{Watt}$$

$$T_{\text{mech}} = 25.04 + 0.5 = 25.54 \text{N.m}$$

$$P_{\text{mech}} = \omega_{\text{mech}} T_{\text{mech}} = 147.76 \times 25.54 = 3773.79 \text{Watt}$$

$$P_{\text{mech}} = P_i (1 - S) \Rightarrow 3773.79 = P_i (1 - 0.04)$$

$$P_i = 3931.588 \text{Watt}$$

$$P_{\text{input}} = P_i + \text{sator losses} = 3931.588 + 320 = 4251.6 \text{Watt}$$

$$P_{\text{input}} = P_{\text{inp}} = \sqrt{3} V_L I_L \cos \phi \quad I_L = 6.88 \text{A}$$

$$\eta\% = \frac{3700}{4251.6} \times 100 = 87\%$$

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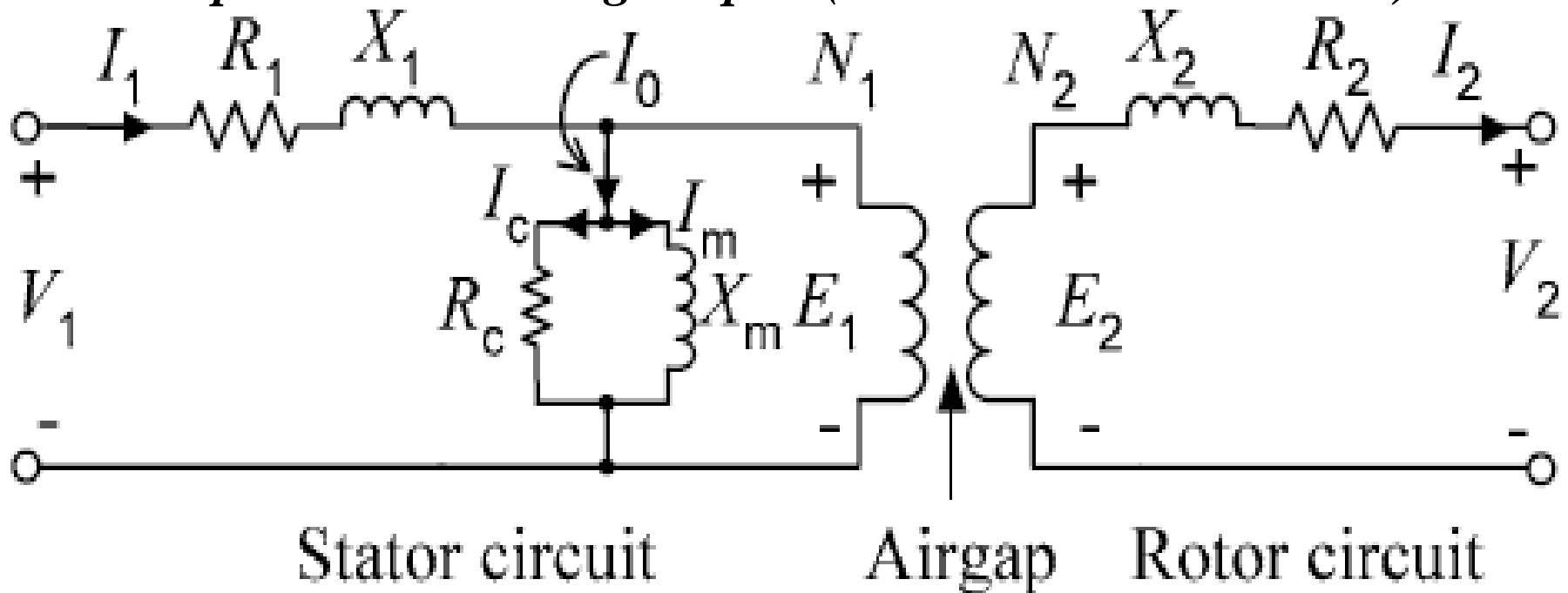
Equivalent Circuit of IM

Conventional equivalent circuit:

Note: 1-*Never use three-phase equivalent circuit. Always use per-phase equivalent circuit.*

2-*The equivalent circuit always bases on the Y connection regardless of the actual connection of the motor.*

Step 1: Rotor winding is open (The rotor will not rotate)



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Equivalent Circuit of IM

Conventional equivalent circuit:

V_1 – stator voltage, per phase ($V_1 = V_{LL} / \sqrt{3}$)

R_1, R_2 – stator and rotor winding resistance

$X_1 = 2\pi f_1 L_1$ – stator leakage reactance

$X_2 = 2\pi f_1 L_2$ – rotor leakage reactance

R_c – resistance representing core loss, per phase

X_m – magnetizing reactance, per phase

N_1, N_2 – effective number of turns of stator and rotor windings.

$E_1 = 4.44 f_1 N_1 \Phi$, where Φ is flux per pole

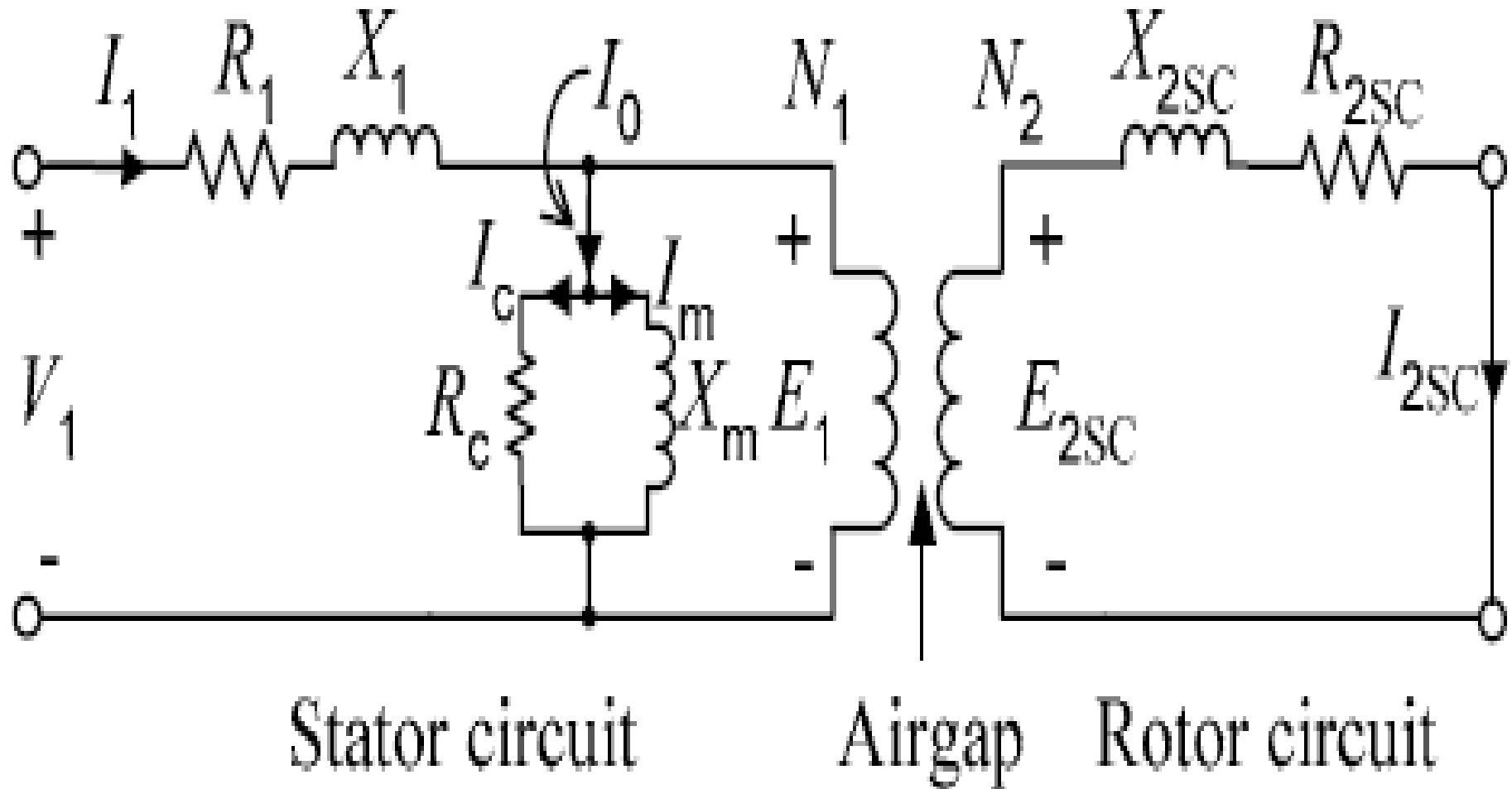
$E_2 = 4.44 f_1 N_2 \Phi$

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Equivalent Circuit of IM

Conventional equivalent circuit:

Step 2: Rotor winding is shorted (Under normal operating conditions, the rotor winding is shorted. The slip is s)



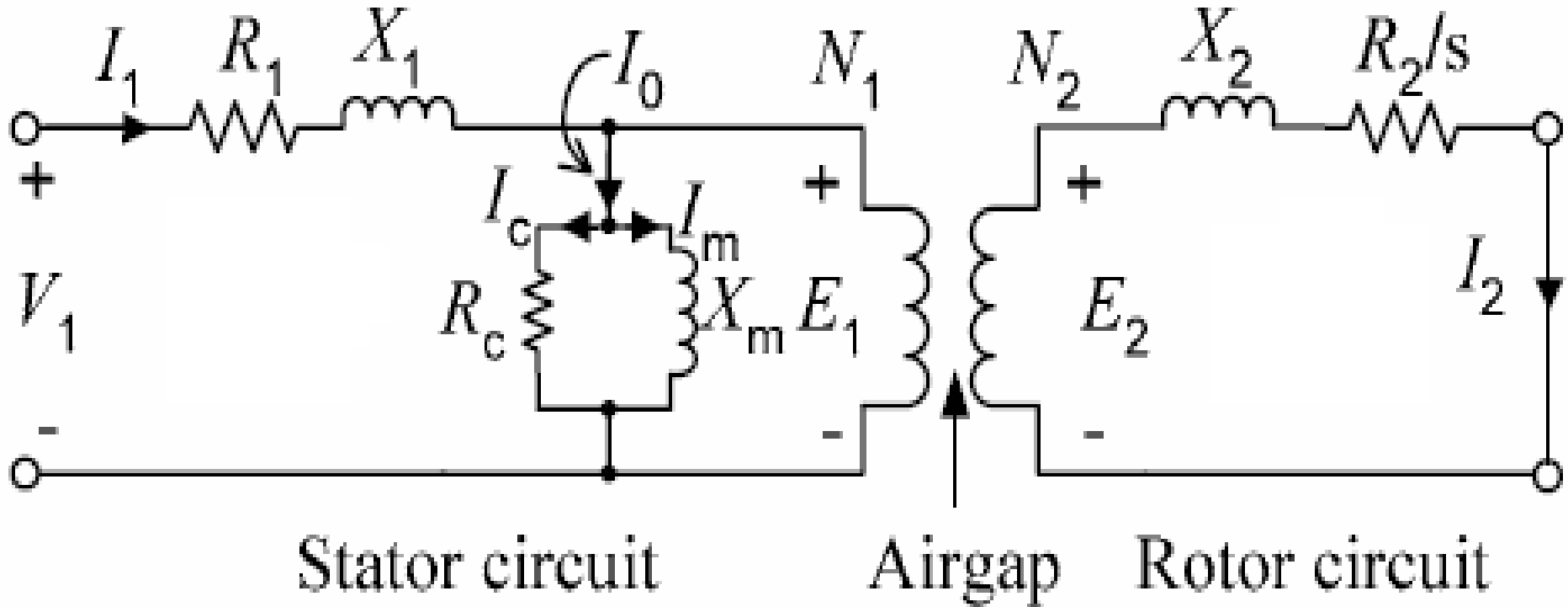
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Equivalent Circuit of IM

Conventional equivalent circuit:

•Note: the frequency of E_2 is $f_r = Sf$ because rotor is rotating. Step 3 : Eliminate f_2

$$I_{2SC} = \frac{E_{2SC}}{R_{2SC} + jX_{2SC}} = \frac{sE_2}{R_2 + jsX_2} = \frac{E_2}{\frac{R_2}{s} + jX_2} = I_2$$

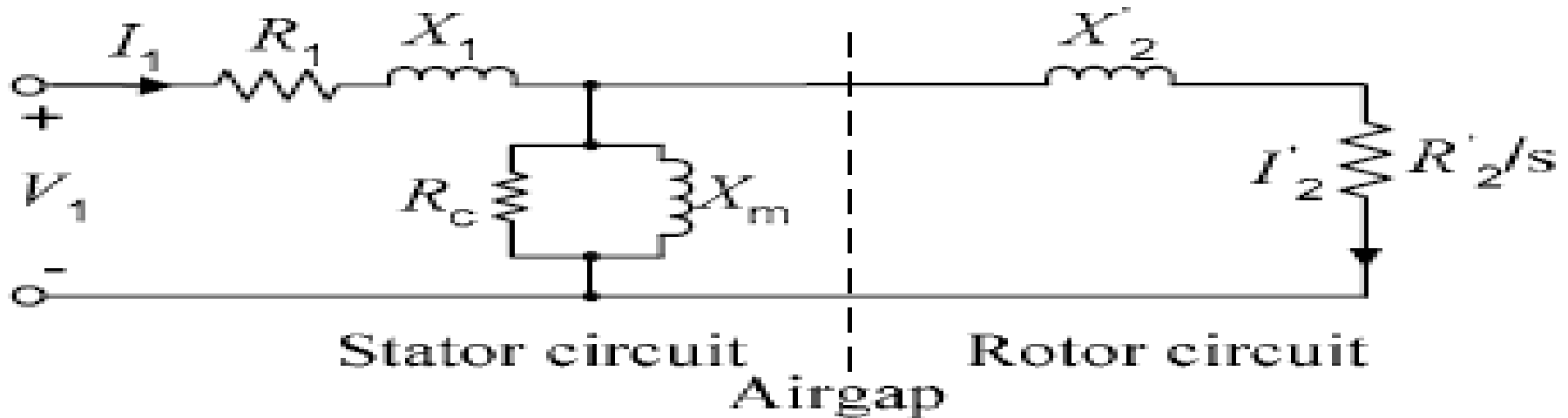


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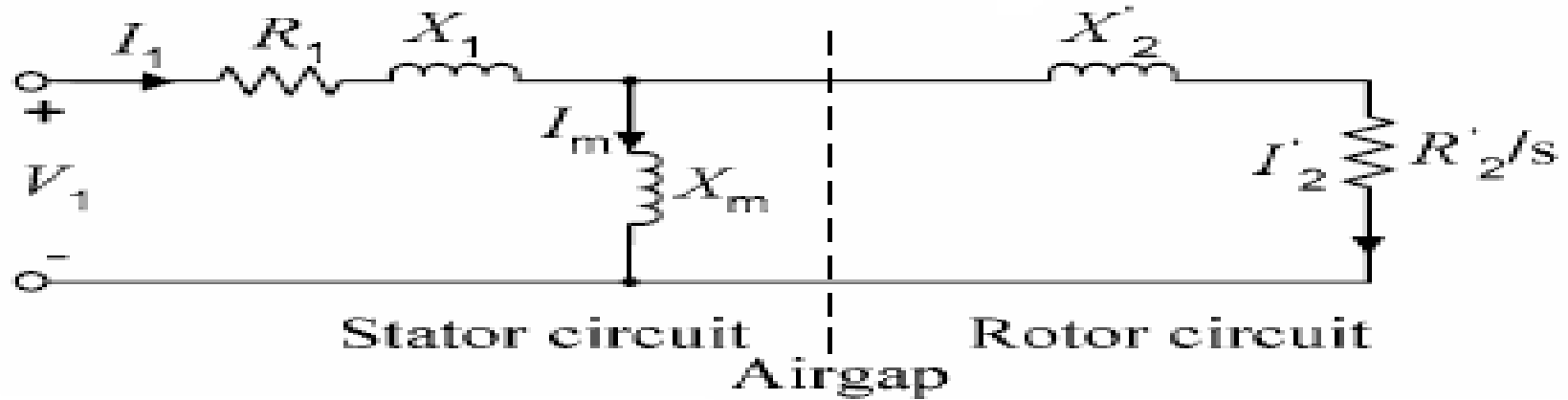
The Exact equivalent circuit:

Step 4 : Referred to the stator side

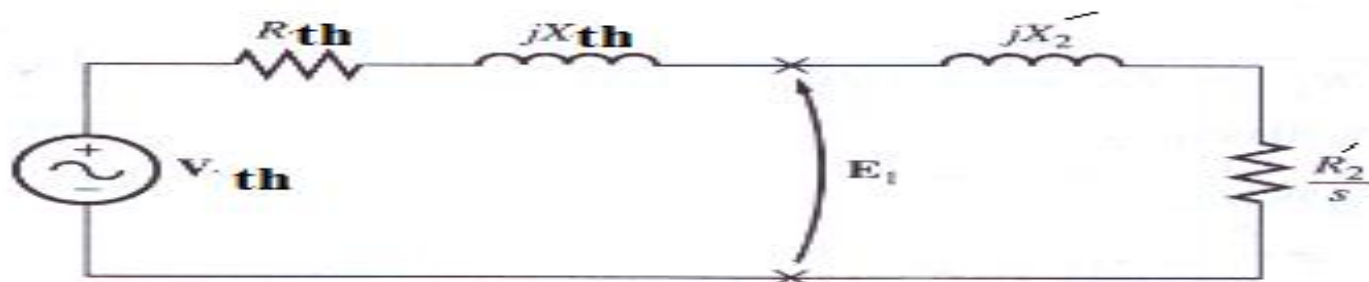
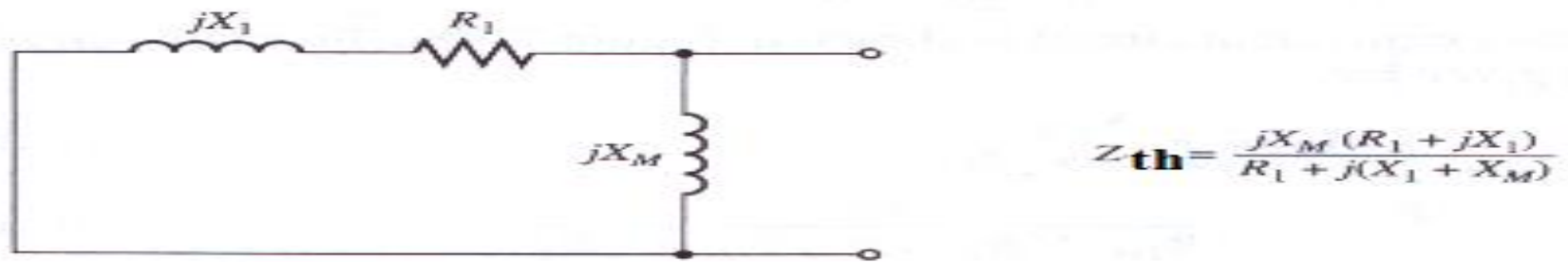
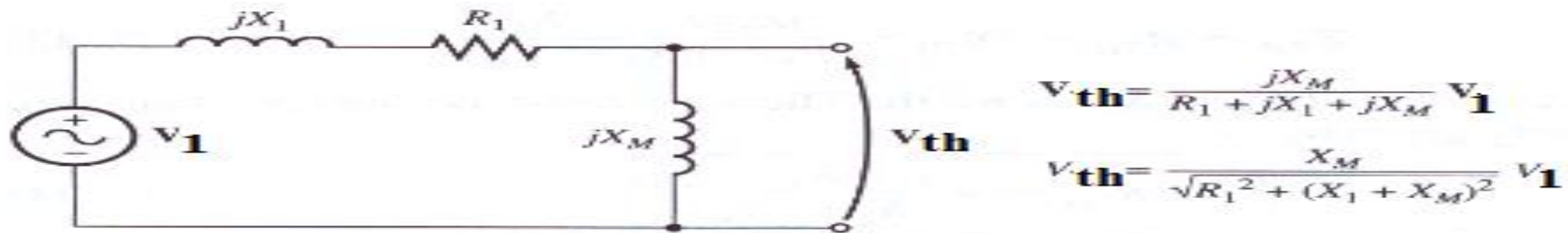


The Exact equivalent circuit:

$$X_2' = a^2 X_2, \quad I_2' = \frac{1}{a} I_2,$$
$$R_2' = a^2 R_2, \quad \text{where } a = \frac{N_1}{N_2}$$



IEEE recommended equivalent circuit



$$V_{th} = V_1 \frac{jX_m}{R_1 + jX_1 + jX_m}, \text{ The magnitude of the Thevenin}$$

voltage V_{th} is

$$V_{th} = V_1 \frac{\overset{\text{IEEE recommended equivalent circuit}}{X_m}}{\sqrt{R_1^2 + (X_1 + X_m)^2}} \approx V_1 \frac{X_m}{X_1 + X_m}$$

(because $X_m \gg (X_1 \ \& \ R_1)$).

$$\text{The Thevenin impedance } Z_{th} = \frac{Z_1 Z_m}{Z_1 + Z_m}$$

$$Z_{th} = R_{th} + j X_{th} = \frac{jX_m (R_1 + jX_1)}{R_1 + jX_1 + jX_m}$$

The Thevenin resistance and reactance are approximately

because $X_m \gg X_1 \ \& \ X_m + X_1 \gg R_1$

$$R_{th} \approx R_1 \left(\frac{X_m}{X_1 + X_m} \right)^2 \ \& \ X_{th} \approx X_1$$

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Equivalent Circuit of IM

Conventional equivalent circuit:

IEEE recommended equivalent circuit

$$\mathbf{I}'_2 = \frac{\mathbf{V}_{th}}{\mathbf{Z}_{th} + \mathbf{Z}_2} = \frac{\mathbf{V}_{th}}{\mathbf{R}_{th} + \mathbf{R}'_2 / s + j\mathbf{X}_{th} + j\mathbf{X}_2}$$

$$\mathbf{I}'_2 = \frac{\mathbf{V}_{th}}{\sqrt{(\mathbf{R}_{th} + \mathbf{R}'_2 / s)^2 + (\mathbf{X}_{th} + \mathbf{X}_2)^2}}$$

$$\mathbf{P}_g = 3(\mathbf{I}'_2)^2 \frac{\mathbf{R}'_2}{s} =$$

$$3 \frac{\mathbf{V}_{th}^2}{(\mathbf{R}_{th} + \mathbf{R}'_2 / s)^2 + (\mathbf{X}_{th} + \mathbf{X}_2)^2} \frac{\mathbf{R}'_2}{s}$$

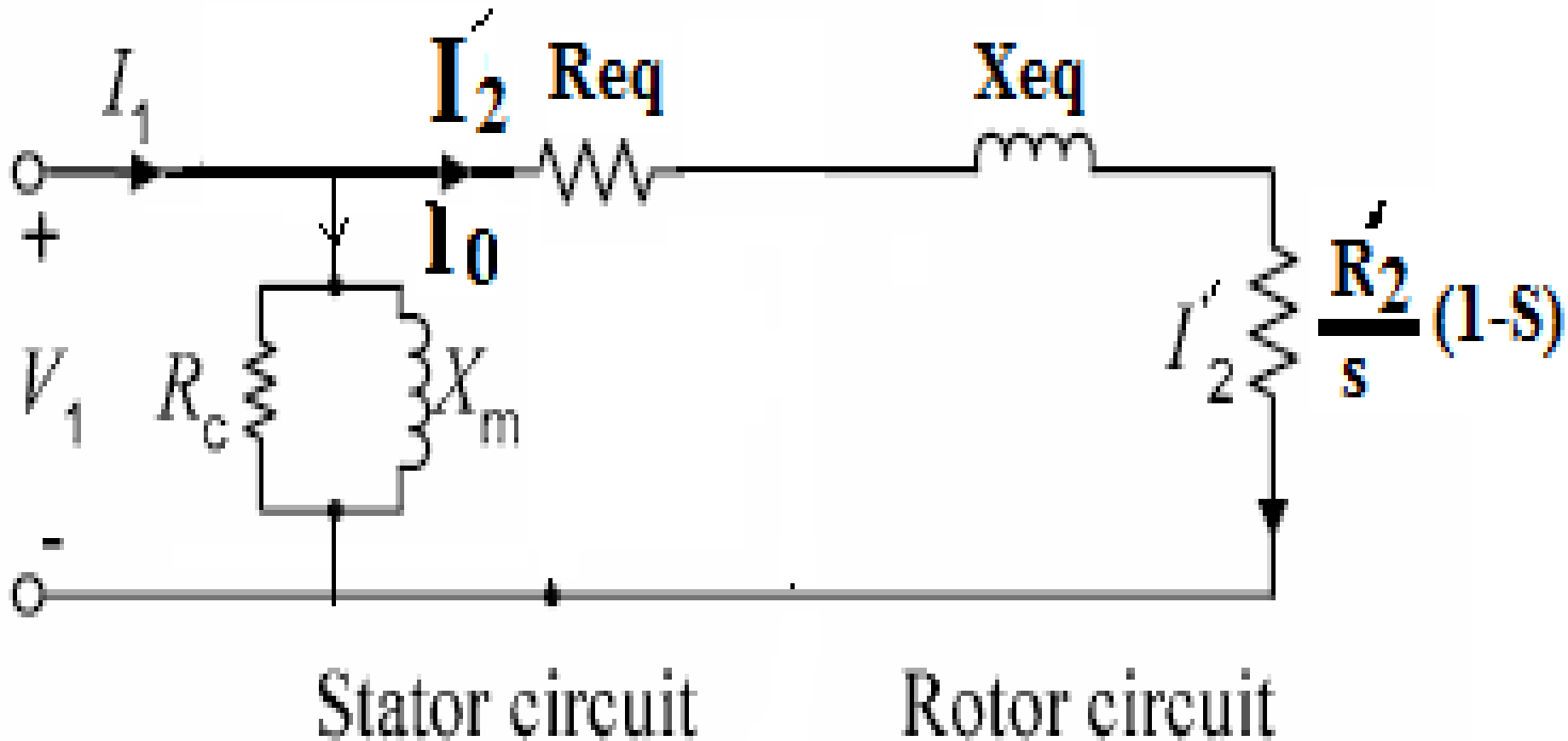
∴ The electromagnetic torque,

$$\mathbf{T}_e = 3 \frac{\mathbf{V}_{th}^2}{\omega_s (\mathbf{R}_{th} + \mathbf{R}'_2 / s)^2 + (\mathbf{X}_{th} + \mathbf{X}_2)^2} \frac{\mathbf{R}'_2}{s}$$

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Approximate equivalent circuit:



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Equivalent Circuit of IM

The Phasor Diagram of Three Phase IM:

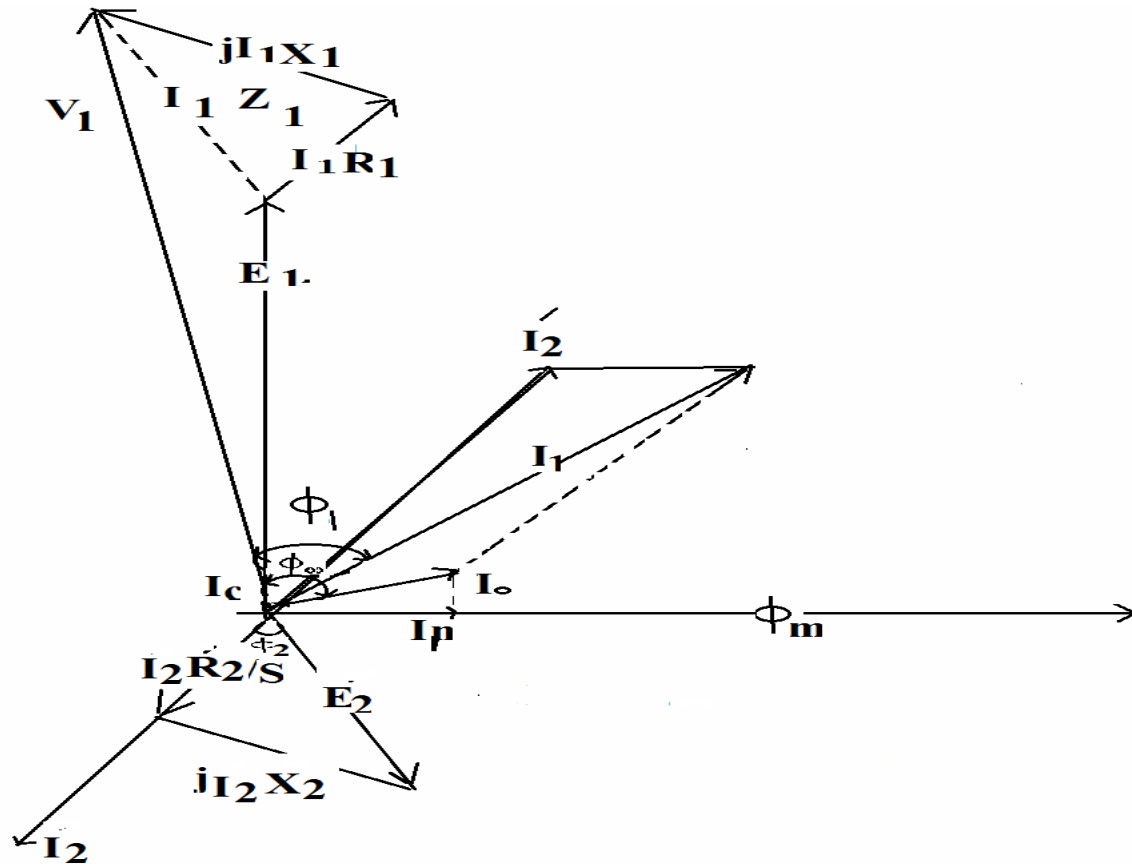
If the rotating magnetic field presented in the air gap considered as a reference, then;

$\phi_m \perp E_1$ Lag. & ϕ_m is in phase with I_m

ϕ_1 is the angle between I_1 & V_1

ϕ_0 is the angle between I_0 & V_1

ϕ_2 is the angle between I_2 & E_2



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Example:

A 415V, delta connected three phase IM has an equivalent circuit of stator impedance $(1+j2) \Omega$ and an equivalent rotor impedance of $(1.2+j1.8) \Omega$. The magnetizing branch impedance of $j50\Omega$ and 500Ω per phase.

Example

1- Use the exact equivalent circuit, find input power, input current, power factor and efficiency at a slip of 0.04. If the mechanical loss equal to 220Watt.

2- Use the approximate equivalent circuit, find input power, input current, power factor and efficiency at a slip of 0.04. If the mechanical loss equal to 220Watt.

3- Use the recommended IEEE standard equivalent circuit, find input power, input current, power factor and efficiency at a slip of 0.04. If the mechanical loss equal to 220Watt.

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Equivalent Circuit of IM

Example

Solution:

$$1- \quad Z_2' = R_2' + jX_2' + R_{mech}' = 30 + j1.8 = 30 \angle 3.4^\circ \Omega$$

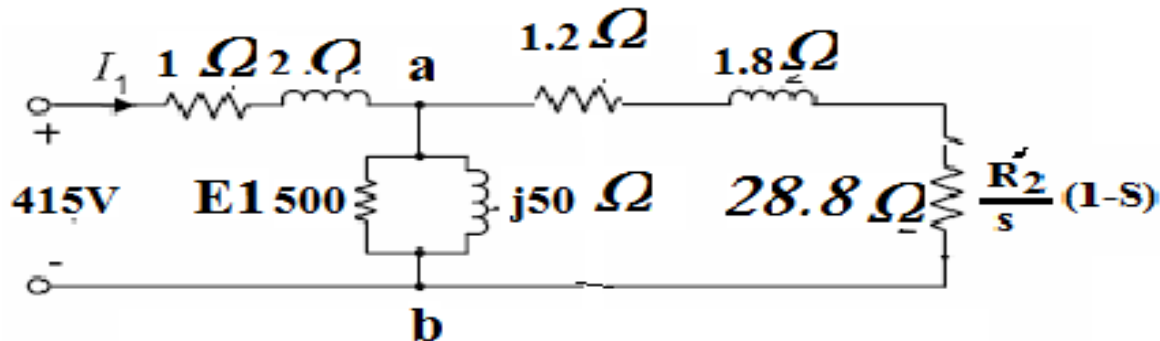
$$Z_M = \frac{500 \times j50}{500 + j50} = 49.75 \angle 84.29^\circ \Omega$$

$$Z_{ab} = Z_2' \parallel Z_M = 24 \angle 31.9^\circ \Omega$$

$$Z_{inp} = Z_{ab} + Z_1 = 25.95 \angle 34.45^\circ \Omega$$

$$I_1 = \frac{V_1}{Z_{inp}} = \frac{415 \angle 0}{25.95 \angle 34.45} = 16 \angle -34.45^\circ \text{ A}$$

$$\cos \phi = \cos 34.45 = 0.82 \text{ lagging}$$



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Example

$$P_{inp} = \sqrt{3} V_L I_L \cos \phi = \sqrt{3} 415 \sqrt{3} 16 0.82$$

$$= 16.426 \text{ kWatt}$$

$$I_2' = I_1 \frac{Z_m}{Z_m + Z_2'} = 12.84 \angle -5.9^\circ \text{ A}$$

$$P_{mech} = 3 (I_2')^2 R_{mech} = 3 (12.84)^2 28.8 = 1411.244 \text{ kWatt}$$

$$P_{out} = P_{mech} - P_{(f+w)} = 14.244 - 0.22 = 14.024 \text{ kWatt}$$

$$\eta\% = \frac{P_{out}}{P_{inp}} \times 100 = \frac{14.024}{16.426} \times 100 = 85.38\%$$

$$1 - \text{Stator cu loss} = 3(I_2')^2 R_1 = 3(16)^2 1 = 768 \text{ Watt}$$

2 - Stator Core loss :

$$\text{First method : } I_o = \frac{Z_2'}{Z_2' + Z_m} \times I_1 = 7.74 \angle -86.78^\circ \text{ A}$$

$$P_{core loss} = 3 |I_o|^2 R_{series \text{ resistance of magnetizing branch}} = 3(7.74)^2 4.95 = 890 \text{ Watt}$$

Second method : by calculating $E_1 = V_1 \angle 0 - I_1 Z_1$

$$E_1 = 415 - 16 \angle -34.45^\circ \times 2.23 \angle 63.43^\circ = 384.24 \angle -2.58^\circ \text{ V}$$

$$I_c = \frac{E_1}{R_c} = \frac{384.2}{500} = 0.768 \text{ A}$$

$$P_{core loss} = 3(I_c)^2 R_c = 886 \text{ Watt}$$

$$\text{cu}_{rotor} = 3(I_2')^2 R_2' = 3(12.84)^2 1.2 = 593.5 \text{ Watt}$$

mechanical loss ($f + w$) = 220 Watt (from quistion)

$$\text{Total loss} = 768 + 890 + 593.5 + 220 = 2471.5 \text{ Watt}$$

$$\text{Also total loss} = P_{inp} - P_{out} = 16426 - 14024 = 2402 \text{ Watt}$$

!—The small difference is due to round off error.

2- The Approximate Equivalent Circuit:

$$\mathbf{Z}_{inout} = (\mathbf{Z}_1 + \mathbf{Z}_2') // \mathbf{Z}_m = 31.23 \angle 7^\circ // 500 // j50$$

$$= 24.16 \angle 35.29^\circ \Omega$$

$$\mathbf{I}_1 = \frac{415 \angle 0^\circ}{24.16 \angle 35.29^\circ} = 17.17 \angle -35.32^\circ \text{ A}$$

$$\cos \phi = \cos 35.32 = 0.816$$

$$\mathbf{P}_{input} = \sqrt{3} 415 \sqrt{3} 17.17 \cdot 0.816 = 17442 \text{ Watt}$$

$$\mathbf{I}_2' = \mathbf{I}_1 \frac{\mathbf{Z}_m}{\mathbf{Z}_m + \mathbf{Z}_1 + \mathbf{Z}_2'} = 13.28 \angle -7^\circ \text{ A}$$

$$\mathbf{P}_{mechanical} = 3 (\mathbf{I}_2')^2 \mathbf{R}_{mech} = 15.237 \text{ kW}$$

$$\mathbf{P}_{out} = 15237 - 220 = 15017 \text{ Watt}$$

$$\eta\% = \frac{15017}{17442} \times 100 = 86\%$$

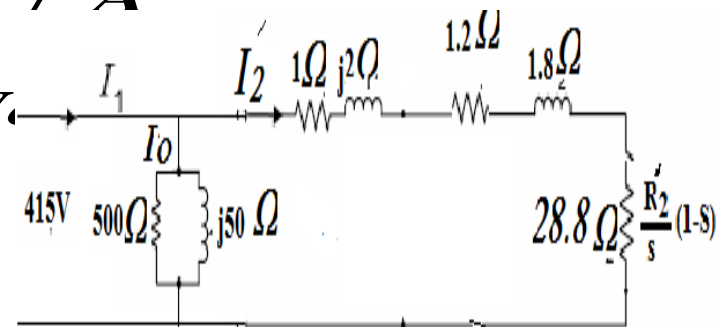
$$\text{Stator copper loss} = 3 (\mathbf{I}_2')^2 \mathbf{R}_1 = 529 \text{ Watt}$$

$$\mathbf{I}_c = \frac{\mathbf{V}_1}{\mathbf{R}_m} = \frac{415}{500} = 0.83 \text{ A}$$

$$\mathbf{P}_{core\ loss} = 3 \mathbf{I}_c^2 \mathbf{R}_c = 1033.35 \text{ Watt} = 3 \frac{\mathbf{V}_1^2}{\mathbf{R}_c}$$

$$\text{cu}_{rotor} = 3 (\mathbf{I}_2')^2 \mathbf{R}_2' = 634.9 \text{ Watt}$$

$$\text{Total loss} = 529 + 1033.35 + 634.9 + 220 = 2417 \text{ Watt}$$



3- Neglecting Rc

$$\mathbf{Z}_{ab} = \mathbf{Z}'_2 \parallel j\mathbf{X}_m = (30 + j1.8) \parallel j50 = 25.06 \angle 33.5^\circ \Omega$$

$$\mathbf{Z}_{inp} = \mathbf{Z}_{ab} + \mathbf{Z}_1 = 27 \angle 35.86^\circ \Omega$$

$$\mathbf{I}_1 = \frac{\mathbf{V}_1}{\mathbf{Z}_{inp}} = \frac{415 \angle 0}{27 \angle 35.86} = 15.37 \angle -33.86^\circ \text{ A}$$

$$\cos \phi = \cos 33.86 = 0.83 \text{ lagging}$$

$$\begin{aligned} P_{inp} &= \sqrt{3} V_L I_L \cos \phi = \sqrt{3} 415 \sqrt{3} 15.3 0.83 \\ &= 15.882 \text{ kWatt} \end{aligned}$$

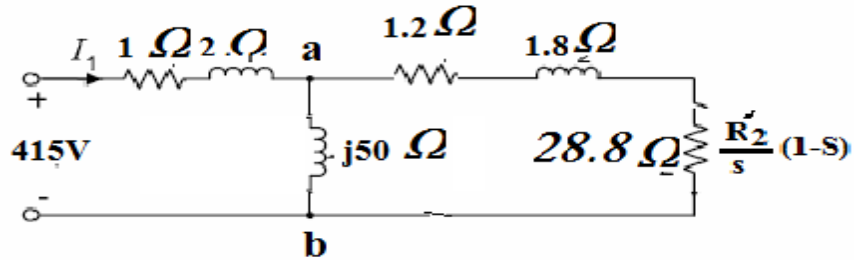
$$\mathbf{I}'_2 = \mathbf{I}_1 \frac{j\mathbf{X}_m}{j\mathbf{X}_m + \mathbf{Z}'_2} = 12.83 \angle -3.04^\circ \text{ A}$$

$$P_{mech} = 3 (I'_2)^2 R_{mech} = 3 (12.83)^2 28.8 = 14.22 \text{ kWatt}$$

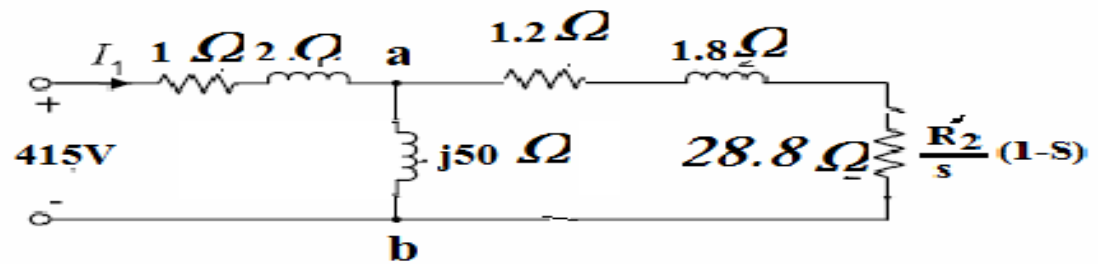
$$P_{out} = P_{mech} - P_{(f+w)} = 14.222 - 0.22 = 14.0 \text{ kWatt}$$

$$\eta \% = \frac{P_{out}}{P_{inp}} \times 100 = \frac{14}{15.882} \times 100 = 88.1 \%$$

$$\text{Stator cu loss} = 3(I'_2)^2 R_1 = 3(15.37)^2 1 = 708.7 \text{ Watt}$$



4- Recommended IEEE



$$V_{th} \approx V_1 \frac{X_m}{X_1 + X_m} \approx 415 \times \frac{50}{2 + 50} = 399V$$

$$R_{th} \approx R_1 \left(\frac{X_m}{X_1 + X_m} \right)^2 \quad \& \quad X_{th} \approx X_1$$

$$R_{th} \approx 1 \times \left(\frac{50}{50 + 2} \right)^2 = 0.924\Omega$$

$$X_{th} = j2\Omega$$

$$I_2' = \frac{V_{th}}{\sqrt{(R_{th} + R_2' / s)^2 + (X_{th} + X_2)^2}} = 12.8A$$

$$P_g = 3(I_2')^2 \frac{R_2'}{s} = 3 \frac{V_{th}^2}{(R_{th} + R_2' / s)^2 + (X_{th} + X_2)^2} \frac{R_2'}{s} = 14.76kWatt$$

$$P_m = P_g - \text{Rotor cu loss}$$

$$= 14.76kW - 3 \times 12.8^2 \times 1.2 = 14170Watt$$

$$P_{out} = 14170 - 220 = 13.95kWatt$$

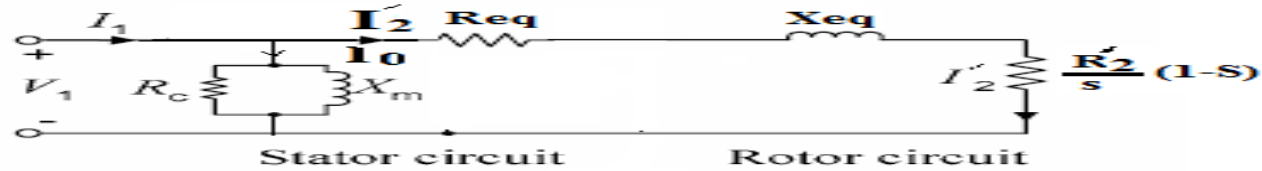
$$\eta\% = \frac{13950}{15882} \times 100 = 87.83\%$$

Comparison between methods

	<u><i>Exact</i></u>	<u><i>Approximate</i></u>	<u><i>IEEE</i></u>	<u><i>Exact without Rm</i></u>
<i>Efficiency %</i>	<i>85.38</i>	<i>86</i>	<i>87.83</i>	<i>88.1</i>
<i>Stator current (A)</i>	<i>16</i>	<i>17.17</i>	<i>15.37</i>	<i>15.37</i>
<i>Power Factor</i>	<i>0.82</i>	<i>0.816</i>	<i>0.83</i>	<i>0.83</i>

Example: A 6-pole, 415V, 50Hz, three phase IM, delta connected has the following data: $Z_1 = (2.2 + j5.8) \Omega = Z_2'$, no-load line current $I_o = 3.54A$, iron loss = 280Watt, mechanical loss = 60Watt. Determine the input current, power factor, out put torque and efficiency if slip is 0.03 using approximate equivalent circuit.

Solution:



$$I_2' = \frac{15 \angle 0}{2.2 + j5.8 + 2.2 / 0.03 + j5.8} = 5.431 \angle -8.73^\circ A$$

$$I_o / \text{phase} = \frac{3.54}{\sqrt{3}} = 2.044 A$$

$$P_{\text{input no-load}} = \sqrt{3} V_o I_o \cos \phi_o \Rightarrow 280 = \sqrt{3} \times 415 \times 3.54 \times \cos \phi_o$$

$\therefore \cos \phi_o = 0.11 \text{ lag} .$

$$I_o = 2.044 \angle \cos^{-1} 0.11 = 2.044 \angle -83.68^\circ A$$

$$I_1 = I_o + I_2' = 6.28 \angle -27^\circ A, \cos \phi = \cos(-27) = 0.89 \text{ lag}$$

$$P_{\text{input}} = \sqrt{3} \times 415 \times \sqrt{3} \times 6.28 \times 0.89 = 6958.6 \text{ Watt}$$

$$P_{\text{out}} = P_i - (cu_s + P_{\text{iron}}) = 3 \times (I_2')^2 \frac{R_2'}{s} (1 - S) - P_{\text{iron}}$$

$$= 6234.4 \text{ Watt}$$

$$T_{\text{out}} = \frac{6234.4}{2\pi \times 970 / 60} = 61.4 \text{ N.m}$$

$$\eta\% = \frac{6234.4}{6958.6} \times 100 = 89.6\%$$